

A Multi Gigabit Clock and Data Recovery Testchip fabricated in 0.18 μ m CMOS

by

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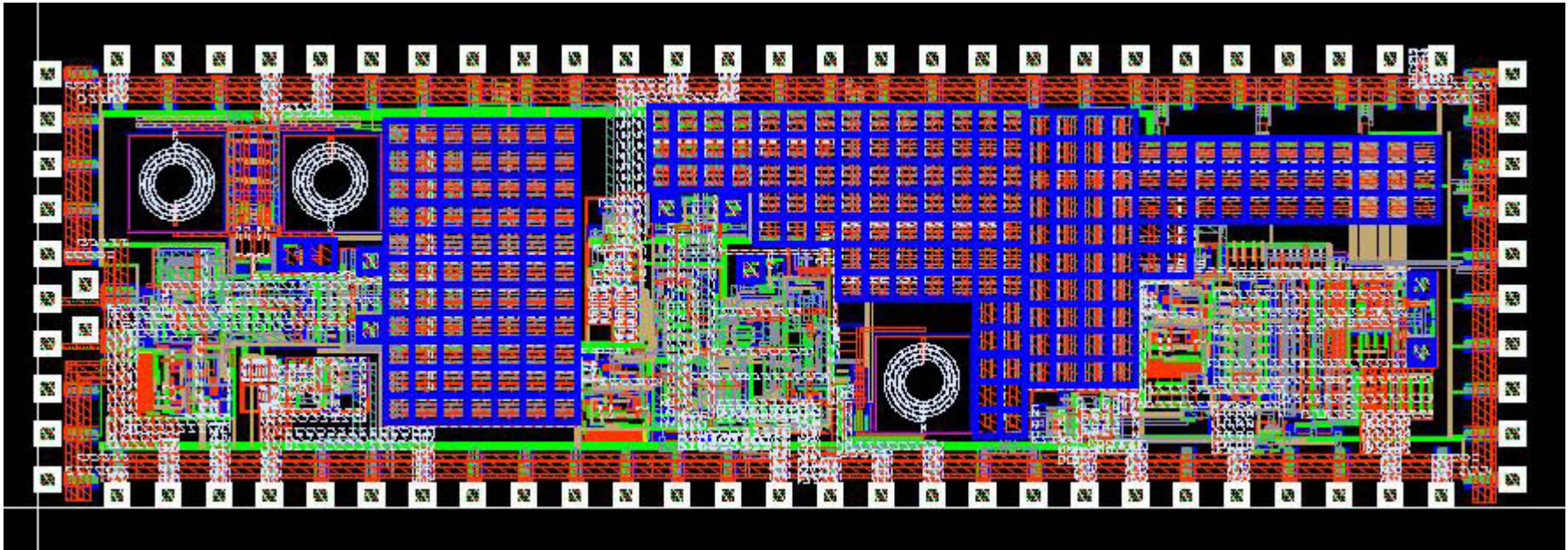
Introduction

- In standard SerDes system, each transceiver has its own local reference clock.
- If clock is distributed through serial data port, clock and data recovery (CDR) circuit has to operate without local reference clock.
- The recovered clock must have sufficient jitter performance to be used as reference clock for local components such as ADC or serializer.
- The conventional half-rate-clock CDR using ring-based oscillator, as using in OASE, do not have sufficient jitter performance.
- Therefore, the feasibilities of low jitter CDR using LC-based oscillator and the improved $1/4^{\text{th}}$ -rate-clock CDR using ring-based oscillator are investigated.

Clock Data Recovery (CDR) Testchip

Technology : UMC 0.18 μm CMOS with RF options

Chip Area : 1525 μm x 4960 μm



CDR-1

- with Quadrature-phase LC-VCO
- full-rate-clock architecture
- current-mode-logic style
- quadricorrelator frequency detector

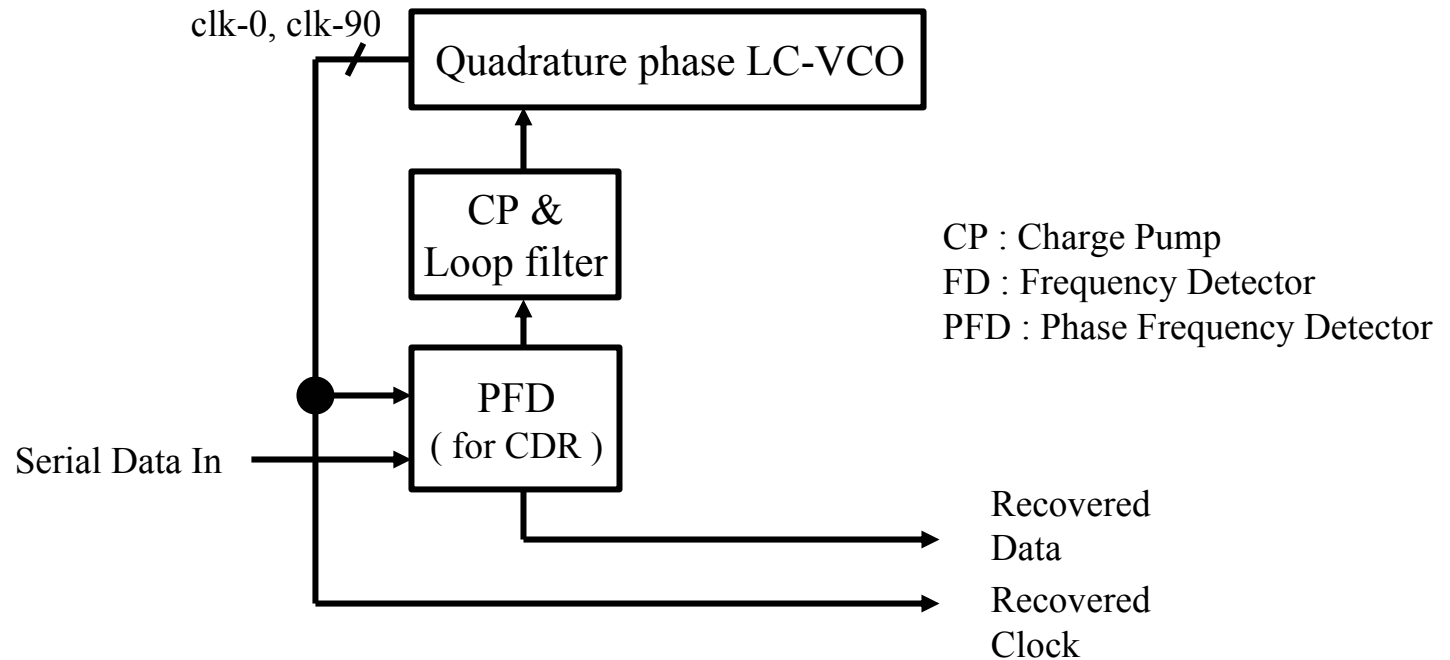
CDR-2

- with LC-VCO and clock delay unit for f-detector
- full-rate-clock architecture
- current-mode-logic
- quadricorrelator frequency detector

CDR-3 with 1:4 DEMUX

- with 8-phase ring oscillator
- 1/4th-rate-clock architecture
- low power CMOS logic style
- parallel architecture
- quadricorrelator frequency detector

Test structure CDR-1 “CDR full-rate Quadraticorrelator FD Structure”



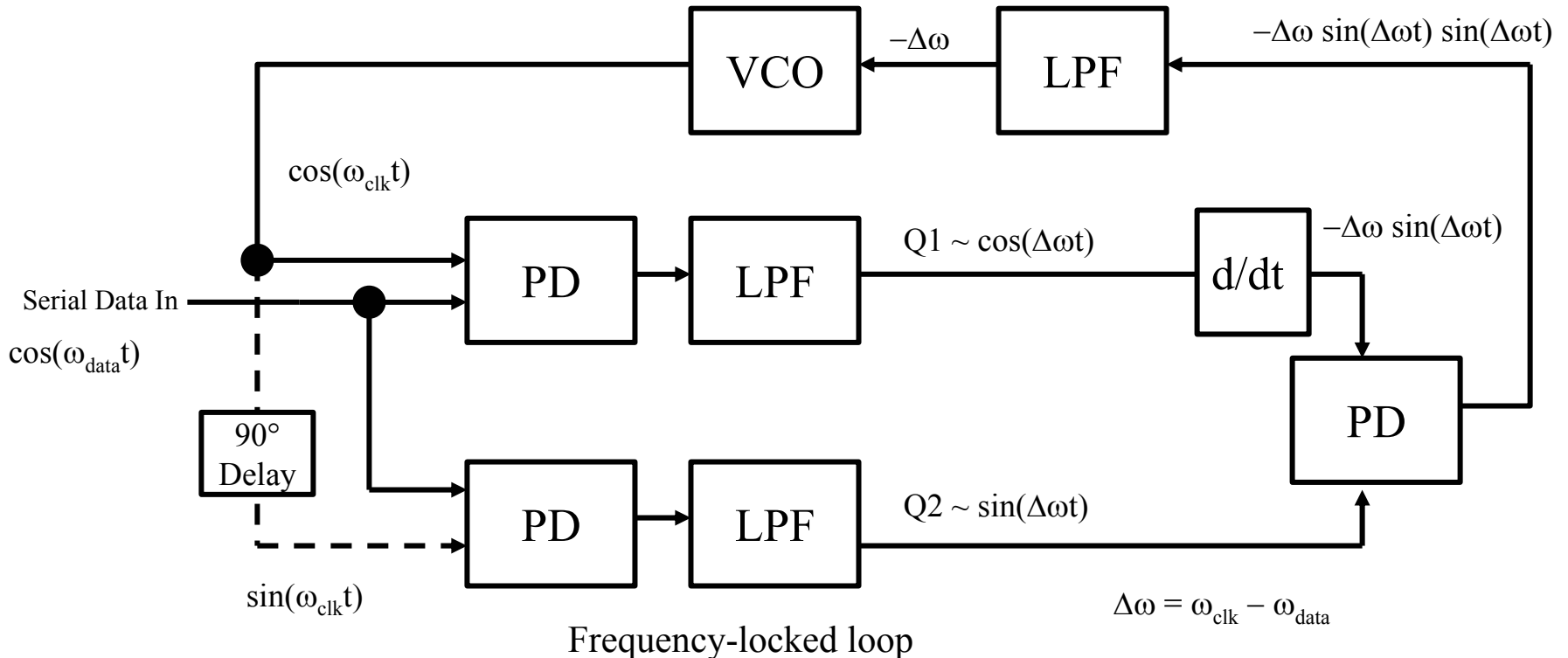
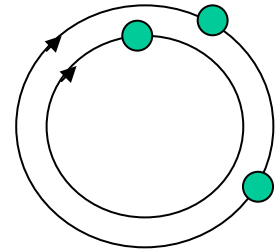
- CDR can operate without local reference clock.
- Quadraticorrelator frequency detector can provides information of frequency difference between VCO clock and serial data steam.
- CDR with FD can have low loop bandwidth and wide pull-in range.

Principle of quadrucorrelator frequency detector

- consists of 2 sets of phase detector with different clock phases
- Q1 and Q2 show the frequency difference.
- Phase difference of Q1 and Q2 shows the sign of frequency difference.

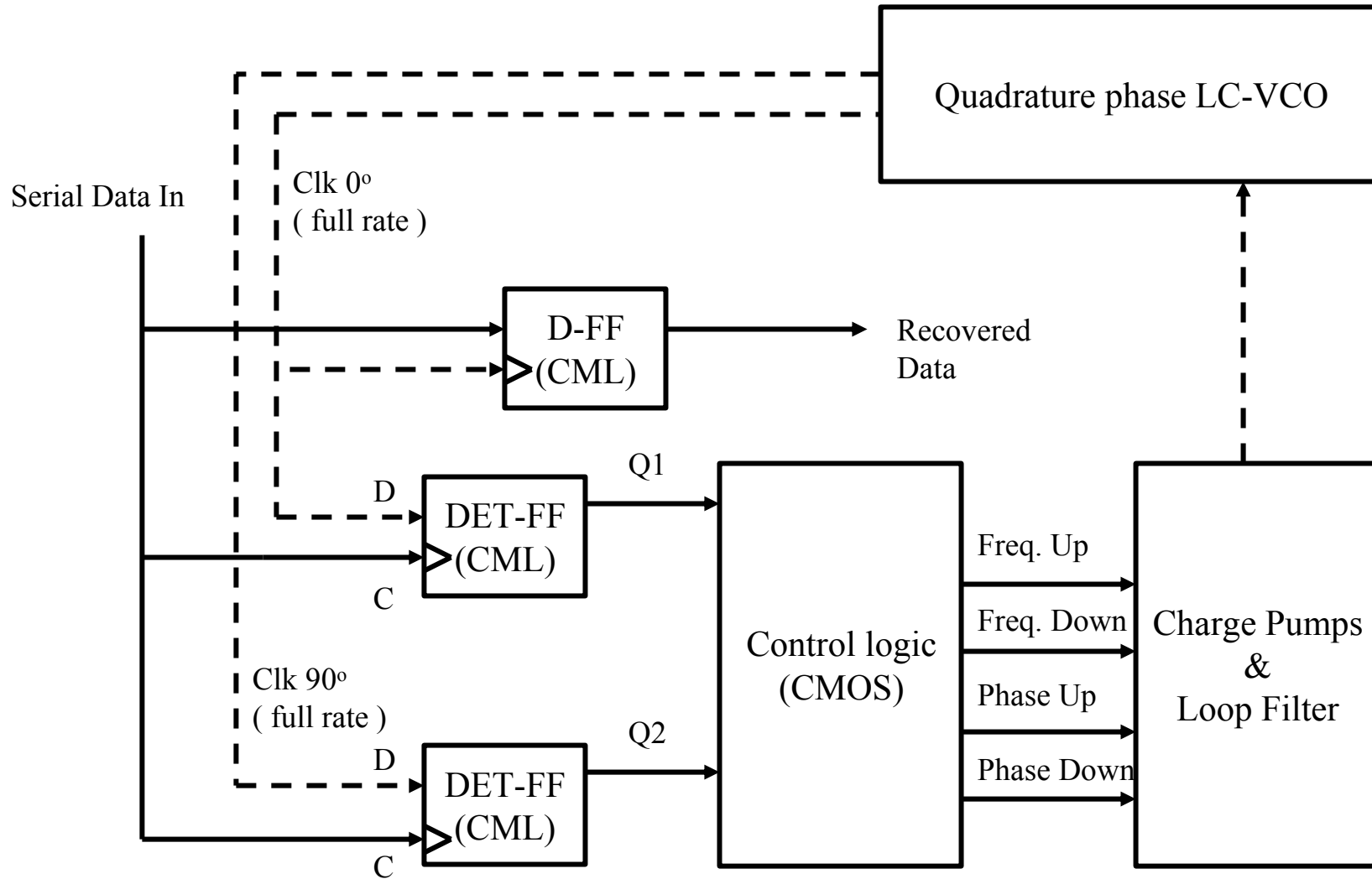
Ex. If $\Delta\omega$ is positive Q1 leads Q2

If $\Delta\omega$ is negative Q2 leads Q1 because $\sin(-\Delta\omega t) = -\sin(\Delta\omega t)$



PD : Phase Detector
 LPF : Low-passed Filter

CDR using digital quadraticorrelator frequency detector

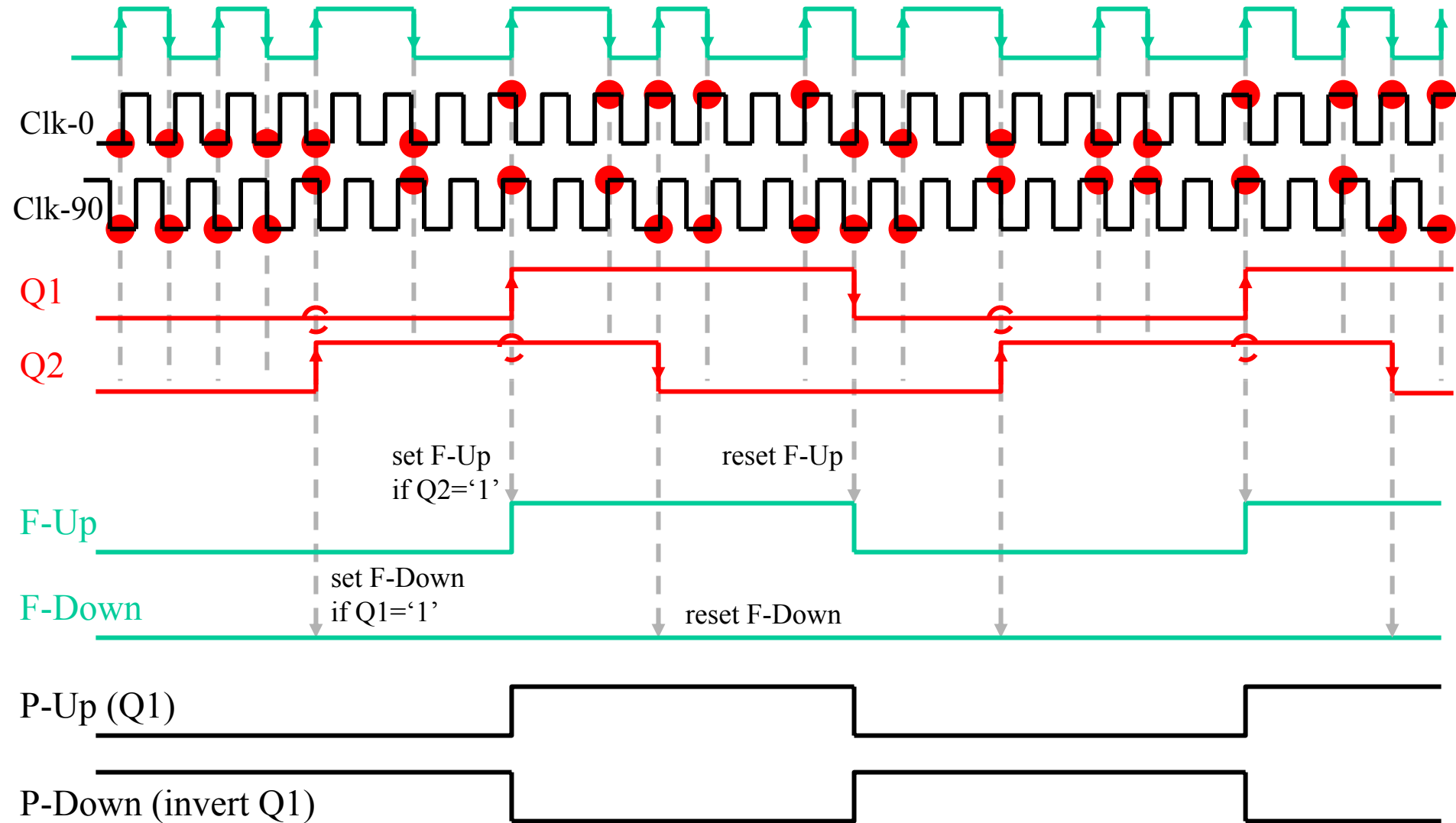


DET FF : Double-edge-triggered Flipflop

PFD Logic of CDR-1: F-VCO too low

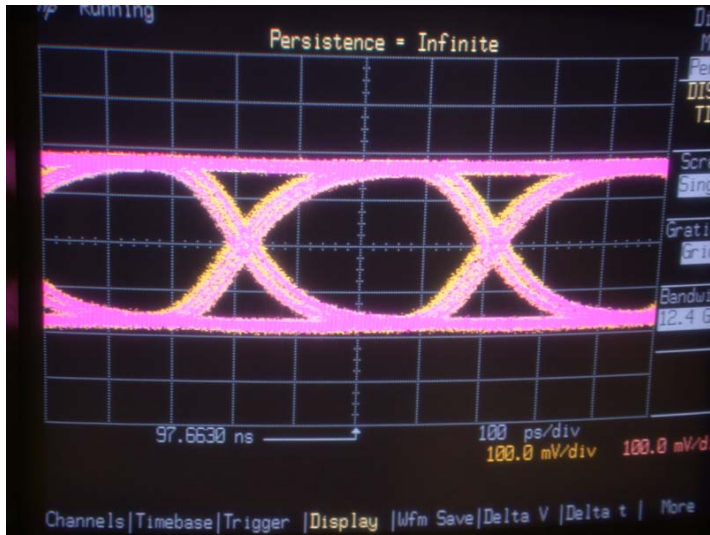
● Sampling point

Serial data

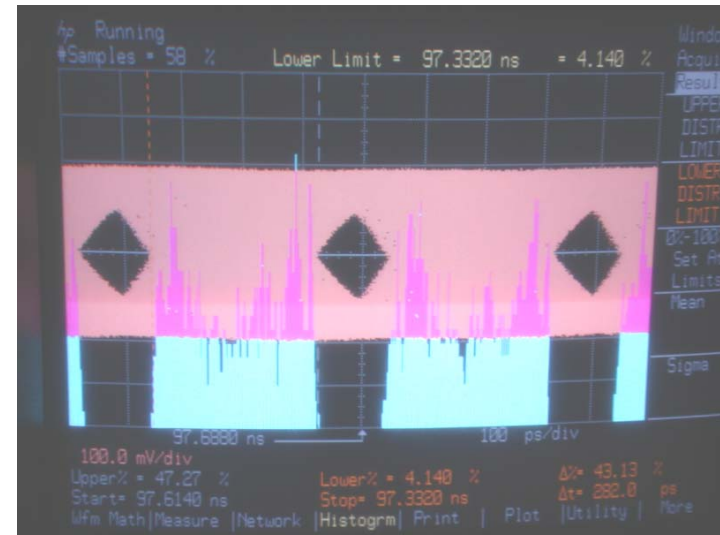


- Both edges of serial data sampling clk-0 and clk-90 to generate Q1 and Q2.
- Q1 and Q2 are used to generate F-Up, F-Down, P-Up and P-Down.

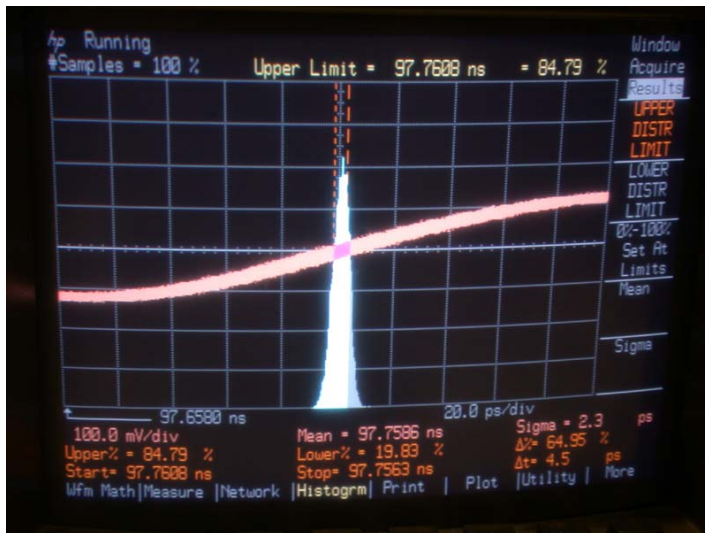
Measurement results : CDR-1



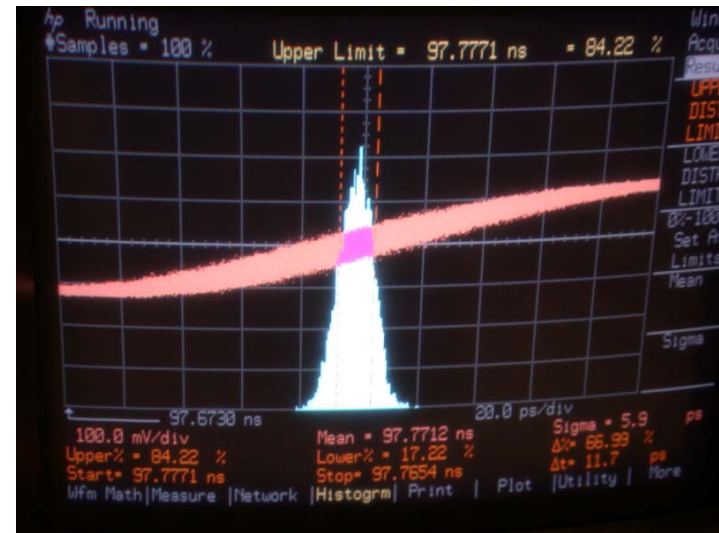
a) Incoming serial data in, 2.5 Gbps, PRBS 27-1



c) Incoming serial data in, 2.5 Gbps, jitter 280 ps,p-p (0.7UI) at 10MHz. modulation

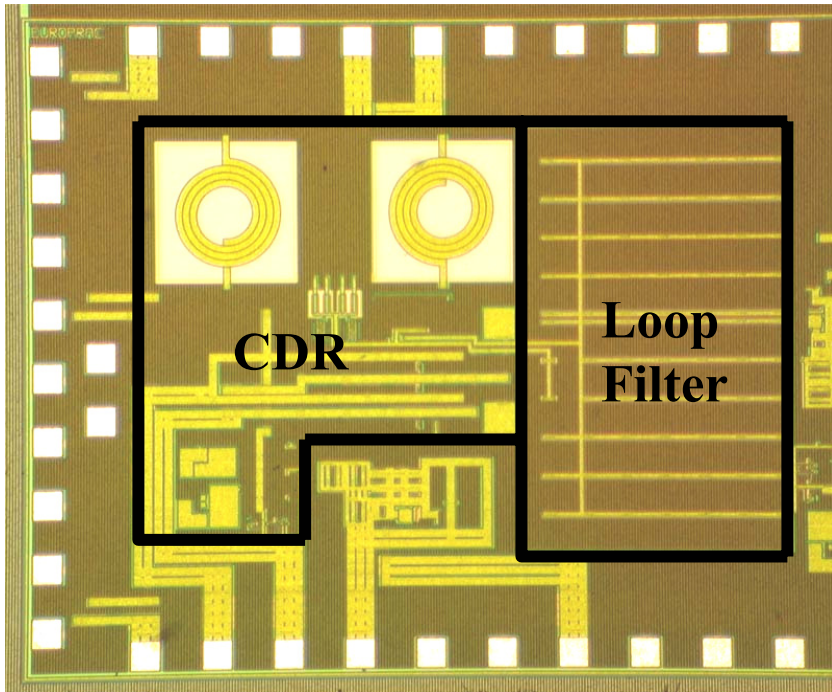


b) Jitter histogram of the recovered clock, 2.5 GHz., jitter 2.3 ps, rms



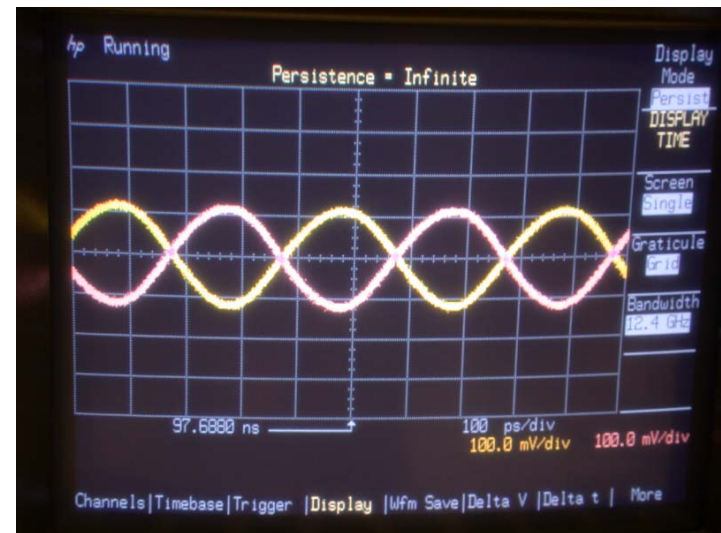
d) Jitter histogram of the recovered clock, 2.5 GHz., jitter 5.9 ps, rms

Measurement results : CDR-1



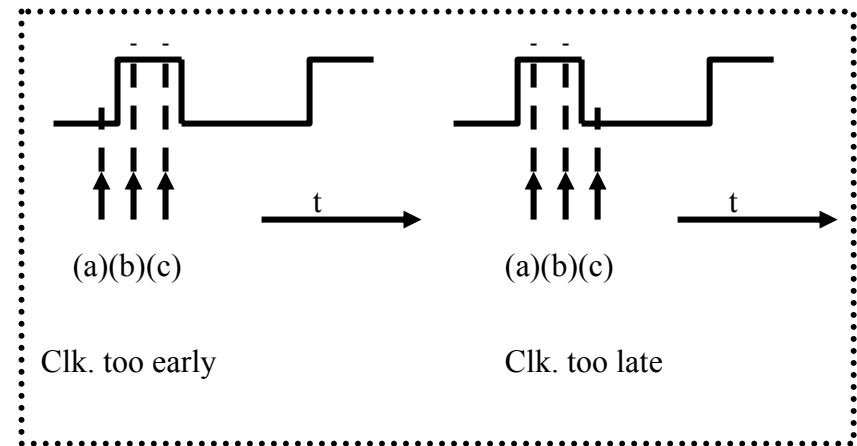
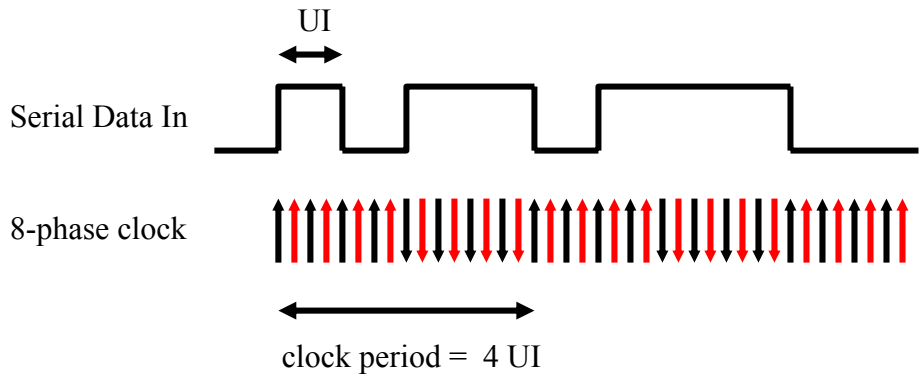
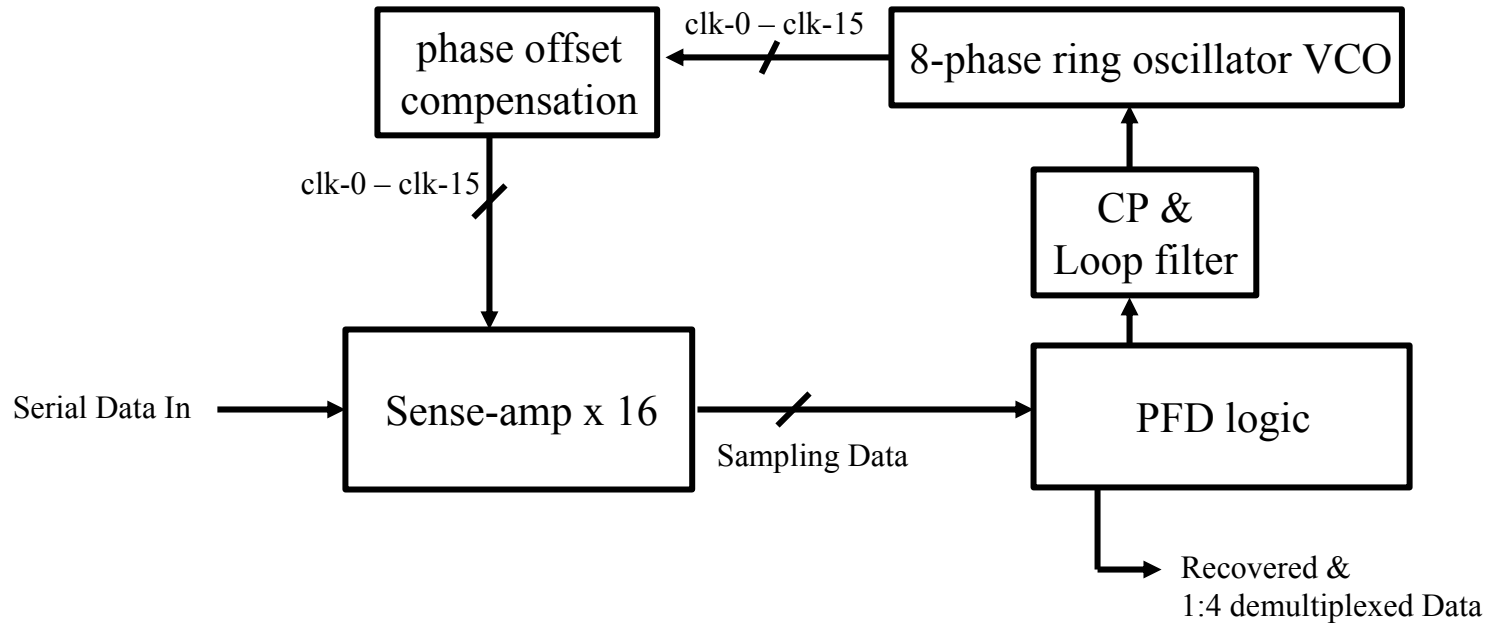
Die Photograph

- Technology 0.18µm CMOS
 - Active area
 - CDR 0.71 mm²
 - Loop Filter 0.69 mm²
 - Power consumption 140 mW at 1.8 Vdd
 - Data rate 2.41 - 2.72 Gbps
 - Loop bandwidth 1.3 MHz.
 - Pull-in range > 100 MHz.
 - Jitter Performance
 - Input Data
 - Output clock
 - 2.5 Gbps, PRBS 2⁷-1 2.3 ps, rms
 - 2.5 Gbps, jitter 280 ps, p-p 5.9 ps, rms
- (0.7UI), modulation at 10 MHz.

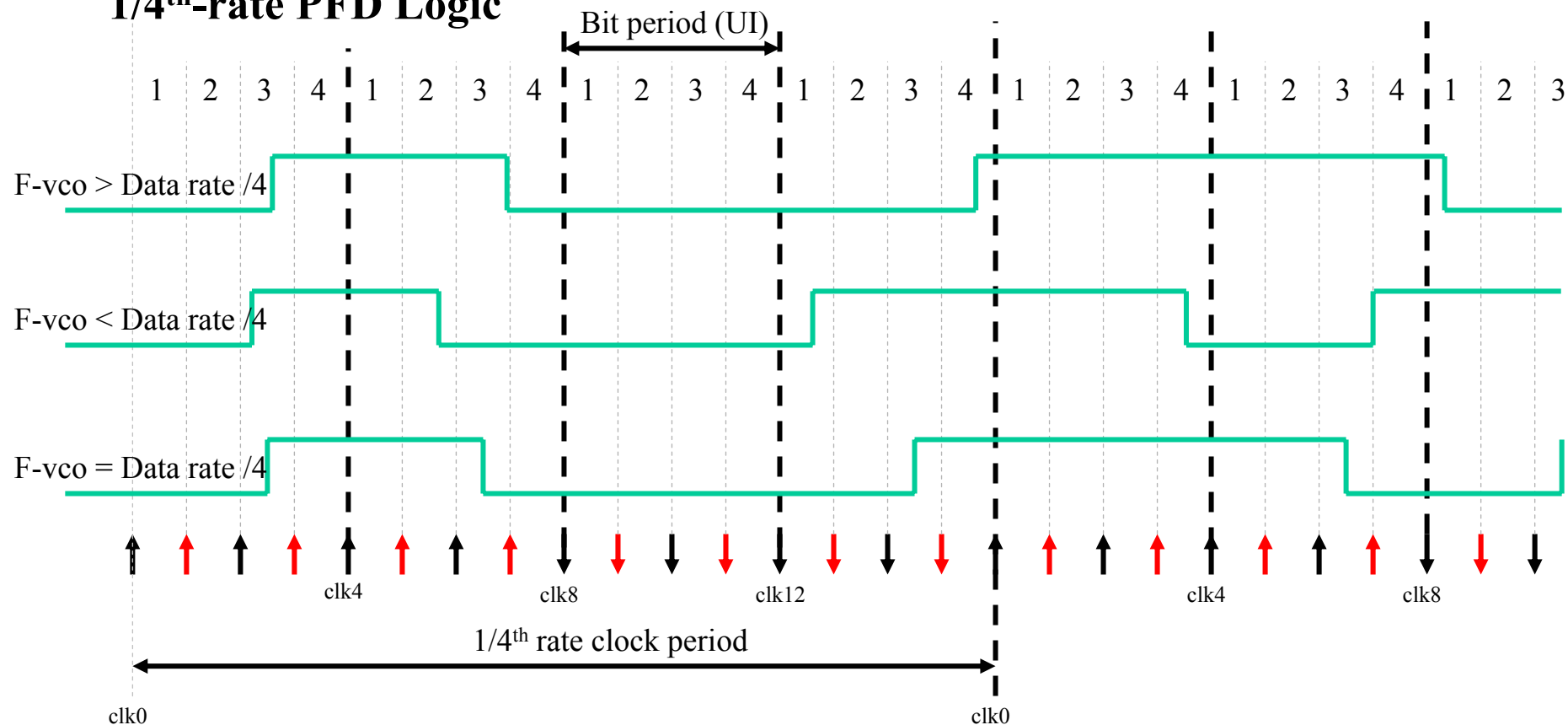


Recovered clock, 2.5 GHz.

Test structure CDR-3 “CDR using 1/4th-rate Quadraticorrelator FD Structure”

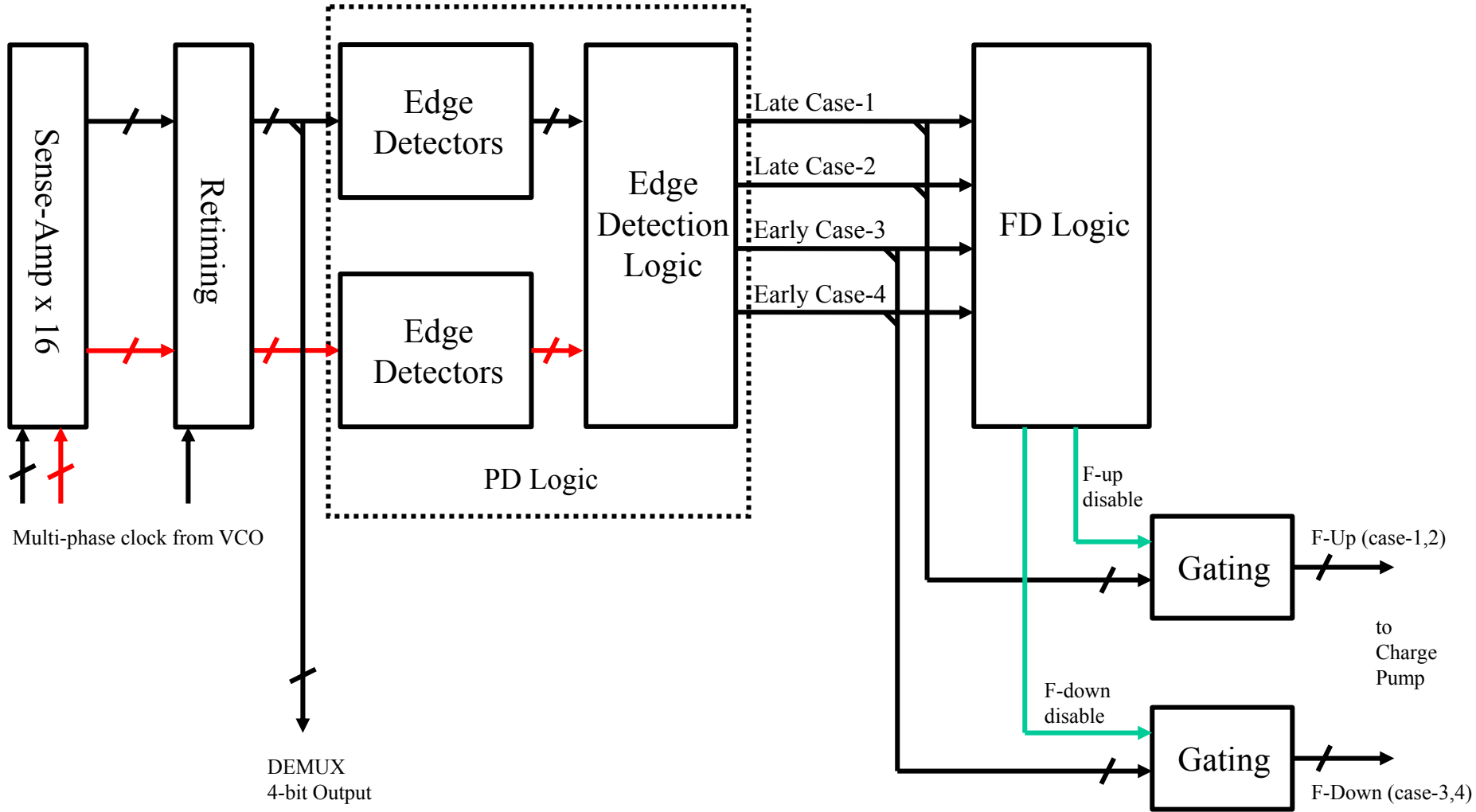


1/4th-rate PFD Logic

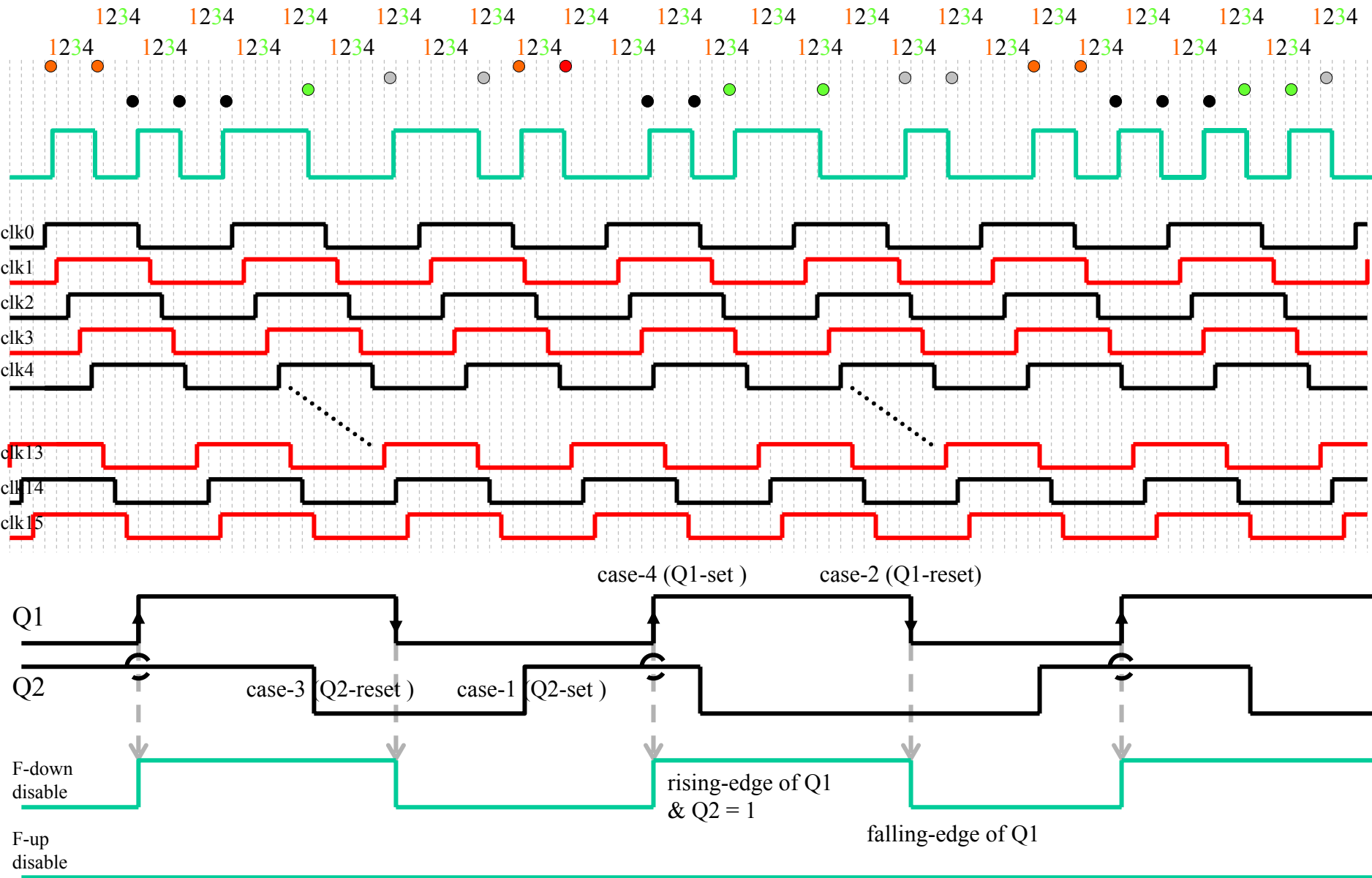


- F-VCO > Data rate /4, data transitions rotate from case-1 -> case-2 -> case-3 -> case-4.
- F-VCO < Data rate /4, data transitions rotate from case-1 -> case-4 -> case-3 -> case-2.
- In lock condition, clk-0 sampling at the middle of data-eye, edges of data are in case-2 or case-3.
 - PD generate "late" signal (f-up) for case-1 and case-2.
 - PD generate "early" signal (f-down) for case-3 and case-4.

1/4th-rate PFD Block Diagram

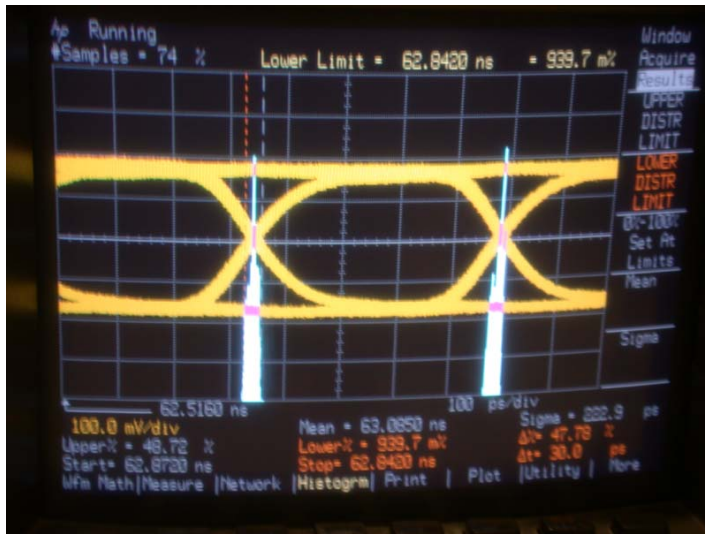


FD Logic of CDR-3: F-VCO too low

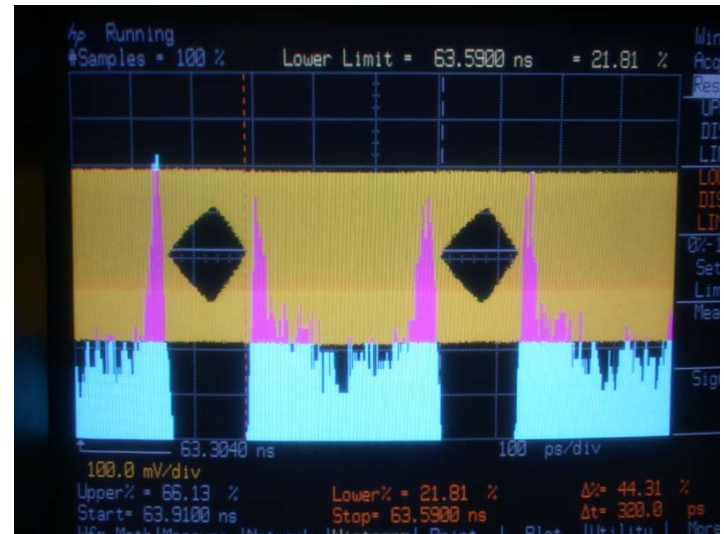


Measurement results : CDR-3

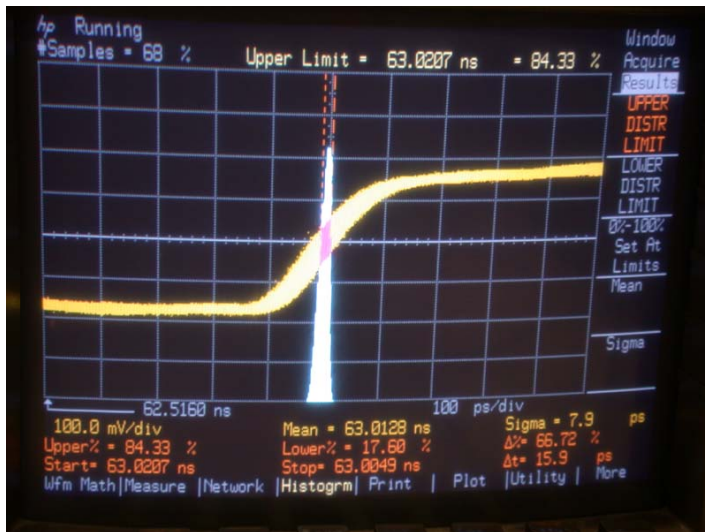
: Jitter Histogram at Incoming Serial Data 2.25 Gbps



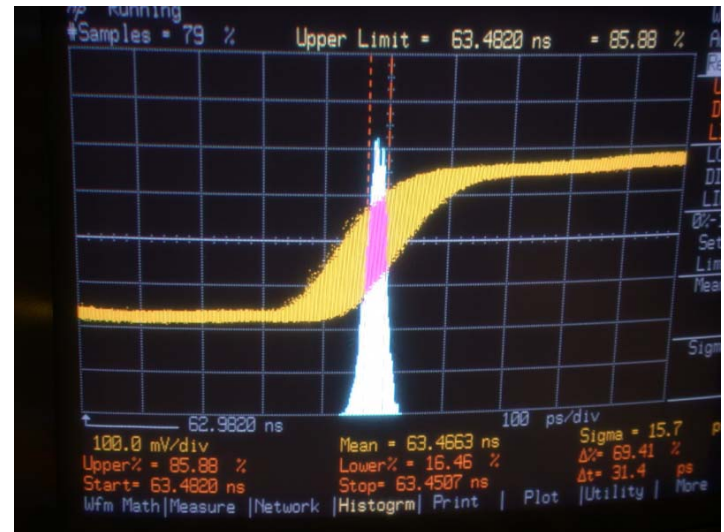
a) Incoming serial data in, jitter 6.6 ps, rms



c) Incoming serial data in, jitter,p-p 311ps (0.7 UI) at 10MHz.

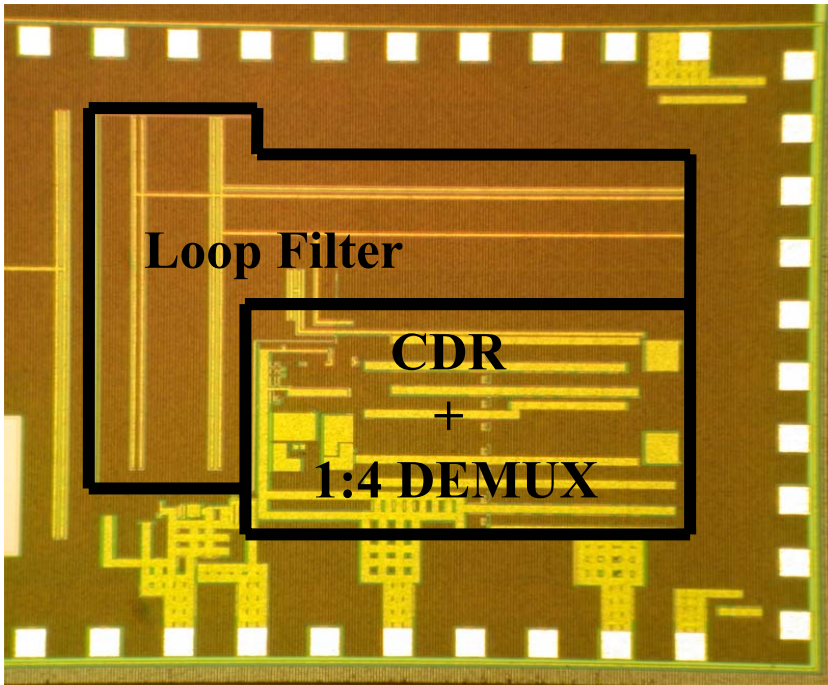


b) Corresponding recovered clock, jitter 7.9 ps, rms



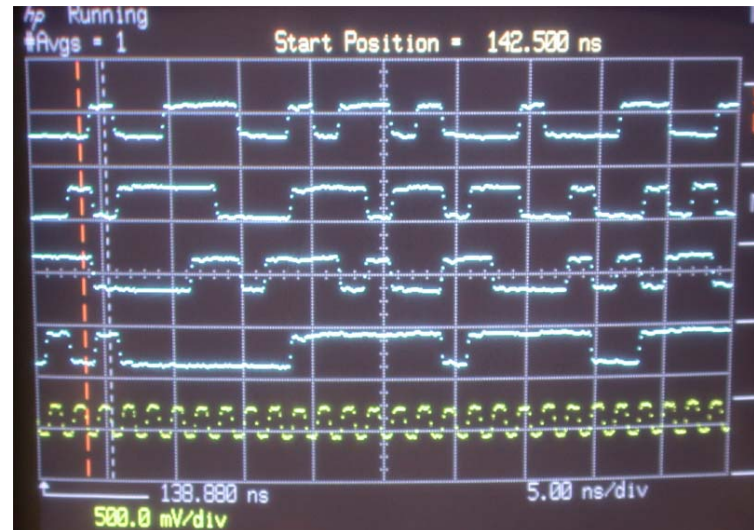
d) Corresponding recovered clock, 15.7 ps, rms

Measurement results : CDR-3



Die Photograph

- Technology 0.18 μ m CMOS
- Active area
 - CDR + DEMUX 0.70 mm²
 - Loop Filter 0.63 mm²
- Power consumption 100 mW at 1.8 Vdd
- Data rate 1 - 2.27 Gbps
- Loop bandwidth 1MHz.
- Pull-in range > 100 MHz.
- jitter Performance
 - Input Data Output clock
 - 2.25 Gbps, PRBS 2⁷-1 7.9 ps, rms
 - 2.25 Gbps, jitter 311 ps, p-p 15.7 ps, rms (0.7UI), modulation at 10 MHz.

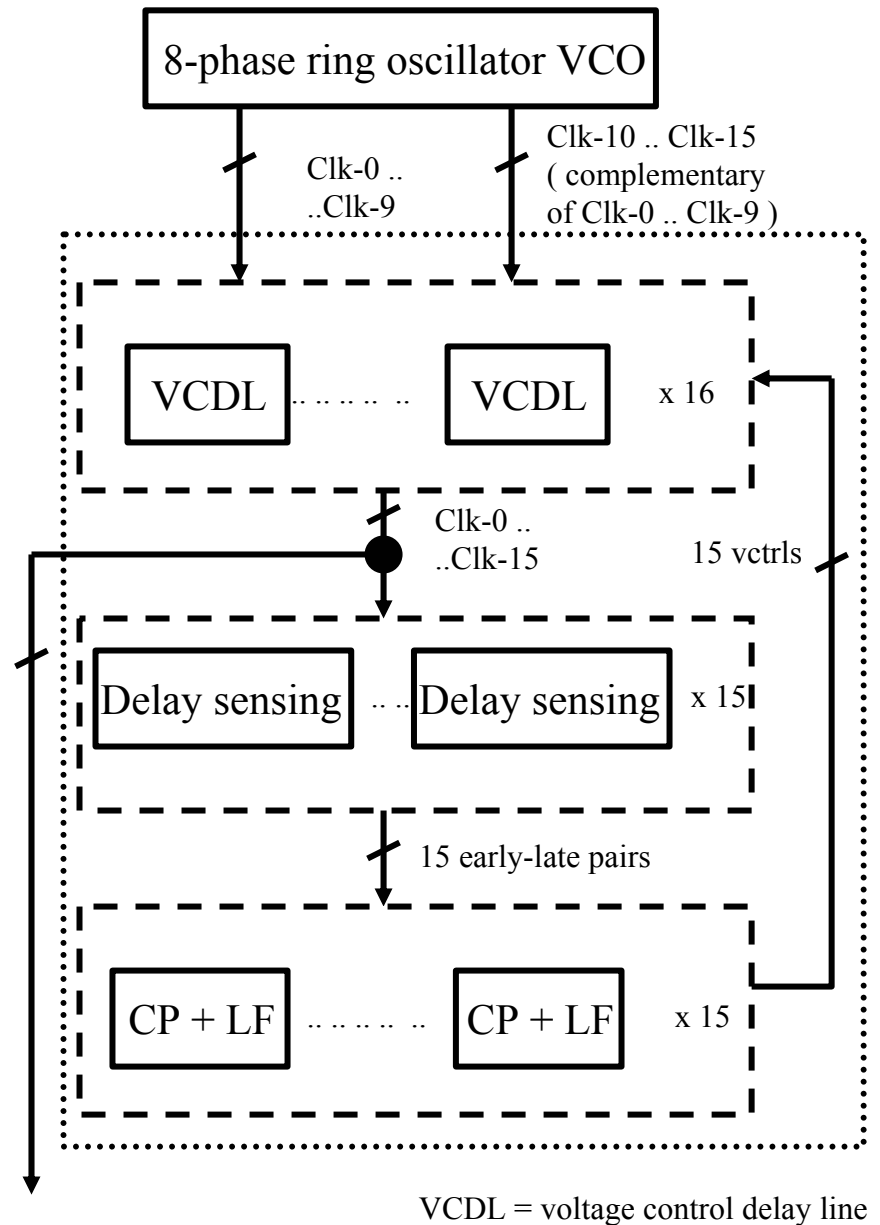
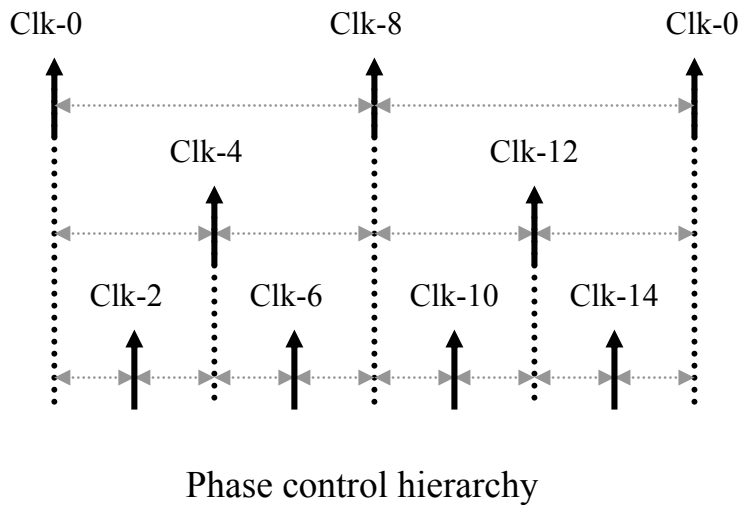


Deserializer Outputs at 562.5 MHz. (2.25 Gbps)

Conclusions

- CDR with Frequency Detector
 - can operate without the need for a local reference clock.
 - low jitter operation and wide pull-in range can be achieved.
- Full-rate CDR architecture,
 - suitable for LC-VCO, very low jitter.
 - phase frequency detector is implemented in CML.
- $1/4^{\text{th}}$ -rate CDR architecture,
 - lower operation frequency, suitable for ring-based oscillator.
 - all logic units can be implemented by CMOS logic, low power.
 - intrinsic 1- to - 4 DEMUX.
 - phase offsets of VCO can be reduced by layout techniques and skew calibration scheme.
- The tested CDRs implemented on $0.18\mu\text{m}$ CMOS Technology has low jitter operation.

Skew Calibration Scheme



Ref: Lin Wu, William C. Black Jr., A Low Jitter Skew-Calibrated Multi-Phase Clock Generator for Time-Interleaved Applications, ISSCC 2001, pp396 – 397.